A CMOS Integrated CC-ISFET Device for Water Quality Monitoring

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Abstract

This study presents a performance analysis of low power CMOS Integrated "Current Conveyor Ion Sensitive Field Effect transistor" (CC-ISFET). The study's main focus is on simulation of power and performance analysis of ISFET device, which is used for water quality monitoring. This approach can improve calibration of device to a fairly wide range without the use of a high speed digital processor. The conventional devices generally used, consume high power and are not stable with temperature and frequency variations for long term monitoring. The conventional device [1] has a drawback of low value of slew rate, high power consumption, and non linear characteristics but in this novel design, the device exhibits a better slew rate, piecewise linear characteristics, and seen consuming low power of the order of 5.7µW. The functionality of the circuit is tested using Tanner simulator version 15 for a 70nm CMOS process model. Very high speed integrated circuit Hardware description language (VHDL) code for the same scheme is simulated on Xilinx ISE 10.1 and various simulation results are obtained. The proposed circuit reduces total power consumption per cycle, increases speed of operation, fairly linear and is simple to implement. This device has a simple architecture, and hence is very suitable for water quality monitoring applications.

Keywords: Slew rate, Calibration, Simulation, Monitoring, Ion Sensitive Field Effect Transistor, Simulation, Current Conveyors, Frequency compensation, Low power.

1. Introduction

Monitoring the pH of water resources for water pollution is a typical and necessary task in today's overdeveloped scenario. The normal pH for surface water systems is 6.5-8.5 and for ground water system 6-8. Water with low pH is acidic, corrosive and contains several toxic materials which are very dangerous for health, but the water having pH more than 8.5 is called hard water which does not contain harmful materials but the long use of such kind of water can cause many problems [2]. With the invention of ISFET [3] there has been a rapid development of pH Measurement Instruments [4]. With the further advancement of semiconductor technology ISFET emerged as a standard device. In spite of the fact that ISFET Sensor has been developed 30 years ago [5], several drawback of ISFET sensor remained unsolved, such as phenomena of fluctuation with time and temperature variations. This causes in drift in the pH values, and result in poor and slow response of the device. The second phenomenon is pH dependent Temperature Coefficient and non linear temperature dependent mobility in MOSFETS of ISFET device. Also it was observed that in ISFET drift rate has an exponential incremental tendency with pH values as well as Temperature. In Urban water supply system, the water quality determining indices such as pH value and turbidity are monitored continuously. When the indices exceed the limiting value, the system will effectively handle the treatment against deterioration ensuring the safety of water. Water is vital for all known forms of life. Many research works have contributed to design water quality measuring devices. But it has always been a challenge to find a precise and accurate device for monitoring the quality of water. The concept of p[H] was first introduced by Danish chemist Soren Peder Lauritz Sorensen at the Carlsberg Laboratory in 1909 and revised to the modern pH in 1924 after it became apparent that electromotive force in cells depended on activity rather than concentration of hydrogen ions. The pH is a measure of the acidity or basicity of an aqueous solution. The use of micro sensors for infield monitoring of environmental parameters is gaining interest due to their advantages over conventional sensors. In the field of micro sensors for environmental applications, Ion Selective Field Effect Transistors (ISFETs) has proved to be of special application. They are particularly helpful for measuring pH and other ions in small volumes and they can be integrated in compact flow cells for continuous measurements and monitoring [6]. Pure water is said to be neutral, with a pH close to 7.0 at 25 °C (77 °F). Solutions with a pH less than seven (7) are said to be acidic and



solutions with a pH greater than seven (7) are basic or alkaline.

1.1 ISFET

An ISFET is an ion-sensitive field-effect transistor which has a property of measuring ion concentrations in solution; when the ion concentration (such as H^+) changes, the current through the transistor will change accordingly. Here, the solution is used as the gate electrode. A voltage between substrate and oxide surfaces arises due to an ions' sheath. The ISFET has the similar structure as that of the MOSFET except that the poly gate of MOSFET is removed from the silicon surface and is replaced with a reference electrode inserted inside the solution, which is directly in contact with the hydrogen ion (H+) sensitive gate electrode [7] The Sub circuit block of ISFET macro model is shown in Figure 1.



Fig. 1. Sub circuit block of ISFET macro model.

At the interface between gate insulator and the solution, there is an electric potential difference that depends on the concentration of H+ of the solution, or so called, pH value. The variation of this potential caused by the pH variation will lead to modulation of the drain current. As a result, the Id-Vgs transfer characteristic of the ISFET, working in triode region, can be observed similar with that of MOSFET:

$$I_{ds} = \frac{\mu C_{ox} W}{L} \left[\left(V_{gs} - V_{th_{isfet}} \right) V_{DS} - \frac{1}{2} \left(V_{ds} \right)^2 \right]$$
(1)

The threshold voltage is only different in case of MOSFET. In ISFET, defining the metal connection of the reference electrode as a remote gate, the threshold voltage is given by:

$$V_{\text{th (ISFET)}} = E_{\text{Ref}} + \Delta \varphi^{1j} - \Psi_{\text{eol}} + \chi^{\text{sol}} + \frac{-\varphi_s}{q} - \frac{Q_{\text{ox}} + Q_{ss}}{C_{\text{ox}}} + \gamma \{2\varphi_{\text{f}}\}^{1/2} + 2\varphi_{\text{f}}$$
(2)

Where E_{Ref} is Potential of reference electrode, $\Delta \phi^{1j}$ is the potential drop between the reference electrode and the solution, which typically has a value of 3mV [8]. Ψ_{eol} is the potential which is pH-independent; it can be viewed as a common-mode input signal for an ISFET interface circuit in any pH buffer solution and can be nullified during system calibration and measurement procedures with a typical value of 50 mV [9]. χ^{sol} is the surface dipole potential of the solvent being independent of pH., the terms in the parentheses are most the same as that of the MOSFET threshold voltage except that of absence of the gate metal function. The other terms in above equation are a group of chemical potential, among which the only chemical input parameter shown has to be a function of solution pH value. This chemical dependent characteristic has already been explained by the Hal and Eijkel's theory which is elaborated using the general accepted site-binding model and the Gouy-Chapman-Stern model. Conventional water quality monitoring applications are made up of voltage mode circuits (VMC) based on op-amps and OTA's. These applications are suffer from low band widths (BW's) arising due to stray and circuit capacitances. Also the need for low voltage, low power circuits makes these circuits not suitable for water quality monitoring as these circuits required the minimum bias voltage depends on the threshold voltage of the MOSFETs. However, with the advancement in the analog VLSI new analog devices are based on currents are developed called current mode circuits (CMC's). These circuits have a significant advantage of low power, low voltages and can operate over wide dynamic range. These circuits, CMC can offer to the designer large bandwidths, greater linearity, wider dynamic range, simple circuitry and low power consumption. Current feedback op-amps (CFOAs), operational floating conveyors (OFCs) and current conveyors (CCs) etc. are popular CMC configuration and most widely used structure among them is CC-II. Hence, one can use the CC-II for the design prospective.

2.0 Current Conveyor

Current Conveyor (CC-II) has proved to be a versatile analog building block that can be used to implement numerous analog processing applications; it was introduced by Sedra and Smith [10]. The current conveyor is a grounded three-port network represented by the black box as shown in Figure 2. The general Current Conveyor (CC) can be represented by the following input-output matrix relation:

| $[I_{Y}]$ | [0 | a | 0] | $[V_{Y}]$ |
|-----------------------|-----|---|----|-----------------------|
| V _X : | = 1 | 0 | 0 | IX |
| $\lfloor I_z \rfloor$ | lo | b | 0] | $\lfloor V_z \rfloor$ |

When a=1, the first generation current conveyor (commonly denoted CCI) is obtained. For a=0, we obtain the second generation current conveyor (commonly denoted CCII). For a=-1 we obtain the third generation current conveyor (commonly denoted CCIII). Usually, $b=\pm 1$. The sign of the b parameter determines the conveyor current transfer polarity. Positive b indicates that the CC has a positive current transfer ratio and is denoted by CCI+, CCII+ or CCIII+ while negative b means that it has a negative current transfer ratio and is denoted by CCI-, CCII-or CCIII-



Fig. 2. Block diagram current conveyor

A well-known basic building block is the second generation current conveyor (CCII). Second generation current conveyors are widely used by analog designers. An important attribute of current conveyor is its ability to convey current between two terminals (X and Z) at vastly different impedance levels.

2.1 CCISFET

For the integrated sensor, the measurement circuit tracks the threshold voltage (or the flat-band voltage) of the ISFET as the electrolyte pH is varied. A practical solution to integrate the sensor with electronics is to view the ISFET sensor as a circuit component in an integrated circuit rather than as an add-on sensor whose output signal is further processed. In this paper, the ISFET is used as one of the input transistors in the differential stage of the current conveyor as shown in Figure 3. The circuit functions as follows: when the ISFET-Current Conveyor is configured as a voltage follower, the output voltage (Vo) is equal to the input voltage (Vin); any difference in threshold voltages and bias currents between the two input transistors at the differential input stage will also appear at the output.



3.0 Simulation and Observations

Simulation of a device for water quality monitoring device using CCISFET involving second generation current conveyors have been carried out on Tanner simulator version 15 for a 70nm CMOS process model shown in Figure 4.The output response of the device with respect to the time i.e transient analysis shown in Figure 6 justify the device is highly linear and having better slew rate than the conventional device . The power results obtained when the device is simulated 70nm technology is shown in the appendix at the end of the paper and it is found that the device consumes the average power of 5.7 μ W watts.



Fig. 4. Circuit diagram of device









Fig.6.Highily linear output waveform



Fig. 7. RTL of CCISFET







Fig. 9. Objects used in the device







| Table 1. Transient analysis | | | |
|-----------------------------|----------------|----------------|--|
| Time <s></s> | v(in1) <v></v> | v(out) <v></v> | |
| 0.000000 e+ 000 | 0.0000e+000 | 0.0000e+000 | |
| 1.250000e-010 | 1.2500e-001 | 9.5606e-002 | |
| 6.786460e-010 | 6.7865e-001 | 5.4378e-001 | |
| 9.554690e-010 | 9.5547e-001 | 7.8071e-001 | |
| 1.093880e-009 | 1.0939e+000 | 8.9976e-001 | |
| 1.316775e-009 | 1.3168e+000 | 1.0922e+000 | |
| 1.600582e-009 | 1.6006e+000 | 1.3394e+000 | |
| 1.922991e-009 | 1.9230e+000 | 1.6229e+000 | |
| 2.303210e-009 | 2.3032e+000 | 1.9608e+000 | |
| 2.758169e-009 | 2.7582e+000 | 2.3703e+000 | |
| 3.296876e-009 | 3.2969e+000 | 2.8694e+000 | |
| 3.931074e-009 | 3.9311e+000 | 3.4780e+000 | |
| 4.678083e-009 | 4.6781e+000 | 4.2192e+000 | |
| 5.000000e-009 | 5.0000e+000 | 4.5821e+000 | |

Table 2. Device process parametersProcess ParametersPower Supply (V)5Load Regulation3.93Line Regulation(m)0.6Current range (µA)1-50Average Power Consumed1.126179e-005(W)Max Power(W)0.574213e-005

Above Table-1 describe the transient behavior of the input with the output and Table-2 describe the various device process parameters including power supply, Load regulation, Line Regulation, Current range, Average power consumed, Max Power.



Fig. 11.Power Comparison Chart



Fig. 12. Component Comparison Chart

Table3: Result analysis of various schemes

| Parameters ▼ | Conventional Device | New Device |
|---|------------------------|-------------------|
| Technology | CMOS | CMOS |
| Power | | |
| supply(VDD, | 5V-0V | 5V-0V |
| GND) | | |
| No. of Mosfets | 27 | 16 |
| Capacitor | 2 | 2 |
| Current Source | 4 | 3 |
| NMOS | 17 | 10 |
| PMOS | 10 | 6 |
| Resistor | 5 | 5 |
| Voltage Source | 4 | 2 |
| Average power dissipation (Watts) | 1.304338e-005 | 1.126179e- 005 |
| Max power (Watts) | 1.635414e-005 | 0.574213e- 005 |
| Min power | 0000000e-000 | 0.131296e- |
| (Watts) | | 005 |
| Stability | Closed loop | Closed loop |
| analysis | Stable | Stable |

Inference from table 3:

For the same technology i.e. CMOS technology deployed for all the four analog IC's and the same power supply we arrive at the following comparative results:

- a. Number of MOSFET's is 27 in Conventional device, 16 in New device, Later deploys 37% of components over 63% used in the previous one shown in Fig. 12.
- b. Number of capacitors remains constant in both techniques.
- c. Number of current sources deployed is 4 in Conventional device and 3 in new device.
- d. Number of n-MOS and p-MOS required are as follows: 17 and 10 for Conventional device, 10 and 6 for new device. From these numbers we can figure out that there is significant saving in terms of Components.
- e. Number of resistors required for both cases are same.
- f. Voltage source require for proper operation of the devices are 4 for Conventional device and 2 in new device.
- g. The average power dissipation in watts is listed for the two cases:
 - 1. Conventional device 1.304338e-005
 - 2. New device 1.126179e-005



Form this we come down to the conclusion that the average power dissipated in the later case is less.

- h. Maximum power dissipated in watts is listed for the four cases:
 - 1. Conventional device 1.635414e-005
 - 2. New device 0.574213e-005

From this we come down to the conclusion that the maximum power consumption in case of new device is 26% as compared to conventional device which consumes 74%. Hence there is total 48% power saving shown in Fig. 11.

4.0 Result and Conclusion

In this novel design, a new device employing second generation current conveyors (CCIIs) is proposed. The second-generation current conveyor introduced is a convenient building block that provides a simplified approach to the design of linear analog systems. It also consumes considerably low power of the order of 5.7 µW compared to [1]. There is significant improvement in the slew rate. The output observed in Fig 6 is highly linear. A significant advantage of the proposed design is it's simple architecture, and low component count a saving of 26% as compared to [1]. Therefore it is very suitable for water quality monitoring applications. VHDL code for the device is simulated on Xilinx ISE 10.1 software and RTL schematic is shown in fig 7. This study may be extended for further improvements in terms of power and size, besides the wiring and layout characteristics level.

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Appendix

| * Synthesis Options Summary * | | |
|----------------------------------|--------------|--|
| Source Parameters | | |
| Input File Name | "Cell3.prj" | |
| Input Format | mixed | |
| Ignore Synthesis Constraint File | : NO | |
| Target Parameters | | |
| Output File Name | : Cell3 | |
| Output Format | : NGC | |
| Target Device | : xc3s100e-5 | |
| vq100 | | |
| Source Options | | |
| Top Module Name | : Cell3 | |
| Automatic FSM Extraction | : YES | |
| FSM Encoding Algorithm | : Auto | |
| FSM Style | : lut | |
| RAM Extraction | : Yes | |
| RAM Style | : Auto | |
| ROM Extraction | : Yes | |
| Mux Style | : Auto | |
| Decoder Extraction | : YES | |
| Priority Encoder Extraction | : YES | |
| Shift Register Extraction | : YES | |
| Logical Shifter Extraction | : YES | |
| XOR Collapsing | : YES | |
| ROM Style | : Auto | |
| Mux Extraction | : YES | |
| Resource Sharing | : YES | |
| Asynchronous To Synchronous | : NO | |
| Multiplier Style | : auto | |
| Automatic Register Balancing | : No | |
| Target Options | | |
| Add IO Buffers | : YES | |
| Global Maximum Fanout | : 500 | |
| Add Generic Clock Buffer(BUFG) | : 24 | |
| Register Duplication | : YES | |



| Slice Packing | : YES |
|-------------------------------|-------------|
| Use Clock Enable | : Yes |
| Use Synchronous Set | : Yes |
| Use Synchronous Reset | : Yes |
| Pack IO Registers into IOBs | : auto |
| Equivalent register Removal | : YES |
| General Options | |
| Optimization Goal | : Speed |
| Optimization Effort | :1 |
| Library Search Order | : Cell3.lso |
| Netlist Hierarchy | : |
| as_optimized | |
| RTL Output | : Yes |
| Global Optimization | : |
| AllClockNets | |
| Read Cores | : YES |
| Hierarchy Separator | :/ |
| Bus Delimiter | :<> |
| Case Specifier | : maintain |
| Slice Utilization Ratio | : 100 |
| BRAM Utilization Ratio | : 100 |
| Verilog 2001 | : YES |
| Auto BRAM Packing | : NO |
| Slice Utilization Ratio Delta | : 5 |
| | |

| * Final Report * | |
|--------------------------------|-------------|
| Final Results | |
| RTL Top Level Output File Name | : Cell3.ngr |
| Top Level Output File Name | : Cell3 |
| Output Format | : NGC |
| Optimization Goal | : Speed |
| Keep Hierarchy | : NO |
| Design Statistics | |
| # IOs | : 4 |
| Cell Usage : | |
| # BELS | : 1 |
| # GND | : 1 |
| # IO Buffers | : 4 |
| # IBUF | : 1 |
| # OBUF | : 3 |
| # Others | : 28 |
| # Capacitor | : 2 |
| # CurrentSource | : 3 |
| # NMOS | : 10 |
| # PMOS | :6 |
| # Resistor | : 5 |
| # VoltageSource | : 2 |

SEDIT: Alter blocks = 0 General options: * Device and node counts: * MOSFETs - 11 **MOSFET** geometries - 4 JFETs - 0 * BJTs - 0 Diodes - 0 * MESFETs - 0 * Capacitors - 1 Resistors - 0 * Inductors - 0 Mutual inductors - 0 Coupled transmission lines - 0 * Transmission lines - 0 Voltage sources - 4 Current sources - 0 VCVS - 0 * VCCS - 0 * CCVS - 0 CCCS - 0 V-control switch - 0 I-control switch - 0 * Macro devices - 0 External C model instances - 0 HDL devices - 0 * Sub circuits - 0 Sub circuit instances - 0

Total nodes - 11

Power Report

*

* Independent nodes - 6

Power Results vdd from time 0 to 1e-008 Average power consumed -> 1.626179e-005 watts Max power 0.574213e-005at time 5e-009 Min power 1.31296e-005 at time 5e-009

Boundary nodes - 5

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